

Control systems

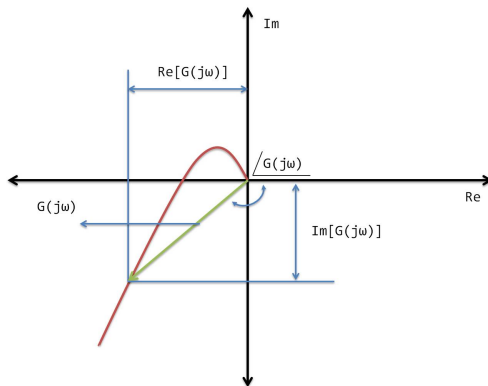
FREQUENCY DOMAIN ANALYSIS: POLAR PLOT

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OUTLINE

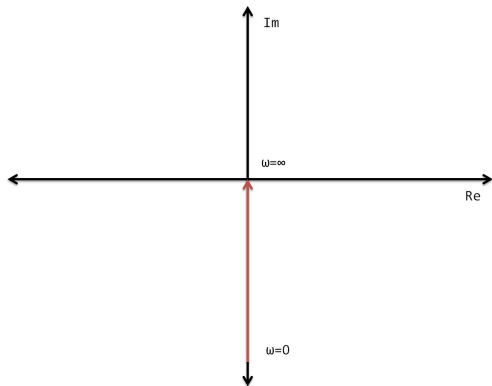
POLAR PLOT

- It is a plot of the magnitude of $G(j\omega)$ versus the phase angle of $G(j\omega)$ on polar coordinates as ω is varied from zero to infinity
- In polar plots a positive (negative) phase angle is measured counterclockwise (clockwise) from the positive real axis



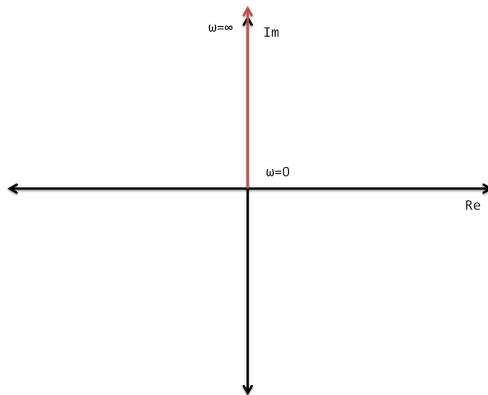
INTEGRAL AND DERIVATIVE FACTORS

- $G(j\omega) = \frac{1}{j\omega} = -j\frac{1}{\omega} = \frac{1}{\omega} \angle -90^\circ$
- The Polar plot of $\frac{1}{j\omega}$ is the negative imaginary axis



INTEGRAL AND DERIVATIVE FACTORS

- The Polar plot of $j\omega$ is the positive imaginary axis



FIRST ORDER FACTORS

- $G(j\omega) = \frac{1}{1+j\omega T} = \frac{1}{\sqrt{1+\omega^2 T^2}} \angle -\tan^{-1}\omega T$
- $G(j0) = 1 \angle 0^\circ$, $G(j\frac{1}{T}) = \frac{1}{\sqrt{2}} \angle 45^\circ$
- $\omega \rightarrow \infty$, $|G(j\omega)| \rightarrow 0$ and $\angle G(j\omega) \rightarrow -90^\circ$

$$G(s) = X + jY$$

$$\frac{1}{1+j\omega T} = \frac{1-j\omega T}{1+\omega^2 T^2}$$

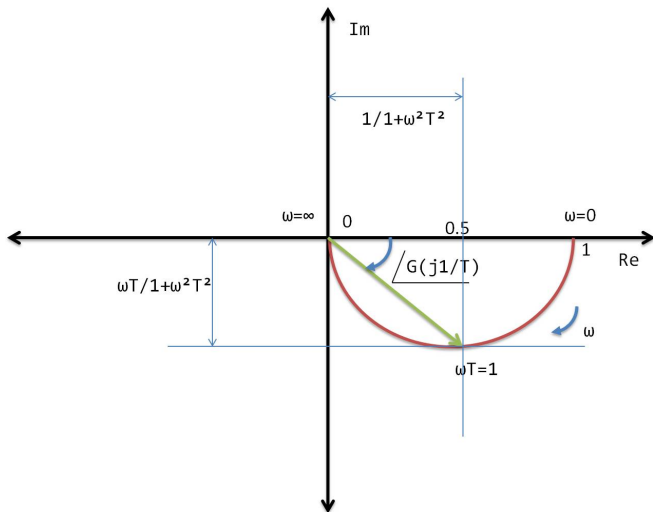
$$G(s) = \frac{1}{1+\omega^2 T^2} - j \frac{\omega T}{1+\omega^2 T^2}$$

$$X = \frac{1}{1+\omega^2 T^2} \quad Y = -\frac{\omega T}{1+\omega^2 T^2}$$

$$\left(X - \frac{1}{2}\right)^2 + Y^2 = \left(\frac{1}{2}\right)^2$$

So the polar plot is a semi circle.

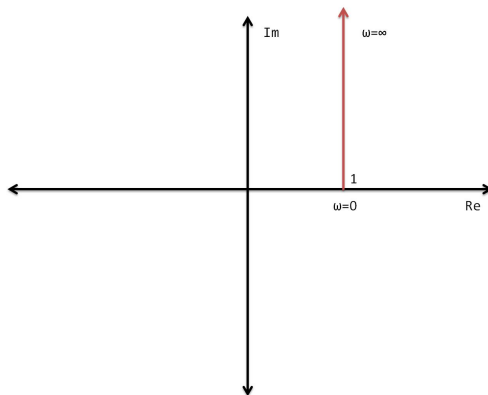
$$G(s) = \frac{1}{1 + j\omega T}$$



FIRST ORDER FACTORS

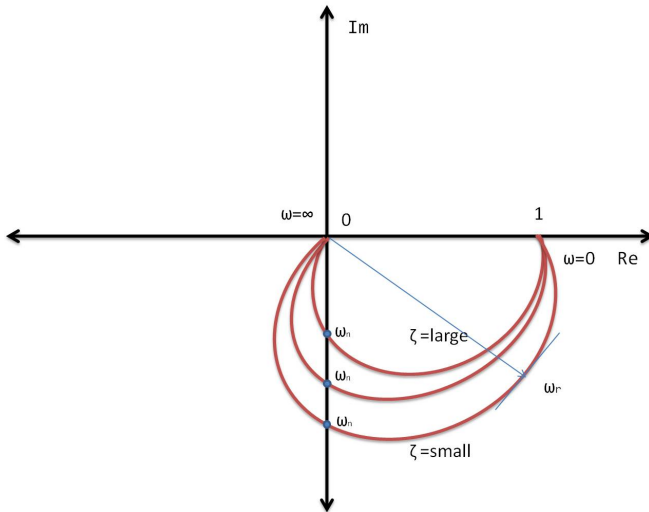
FIRST ORDER FACTORS

- $G(j\omega) = 1 + j\omega T$
- It is simply the upper half of the straight line passing through point $(1, 0)$ in the complex plane and parallel to the imaginary axis



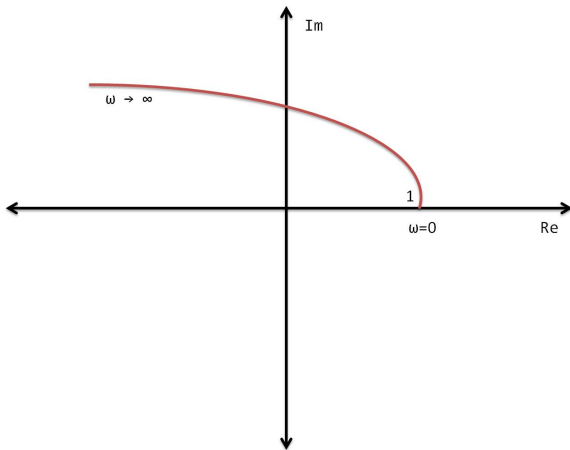
QUADRATIC FACTORS

- $G(j\omega) = \frac{1}{1+2\zeta(j\frac{\omega}{\omega_n})+(j\frac{\omega}{\omega_n})^2}$
- For $\zeta > 0$
- $\lim_{\omega \rightarrow 0} G(j\omega) = 1\angle 0$
- $\lim_{\omega \rightarrow \infty} G(j\omega) = 0\angle -180$
- The polar plot of this sinusoidal transfer function starts at $1\angle 0$ and ends at $0\angle -180$ as ω increases from zero to infinity
- The high-frequency portion of $G(j\omega)$ is tangent to the negative real axis.
- For the underdamped case $\omega = \omega_n$, the phase angle is -90
- In the polar plot, the frequency point whose distance from the origin is maximum corresponds to the resonant frequency ω_r .



QUADRATIC FACTORS

- $G(j\omega) = 1 + 2\zeta(j\frac{\omega}{\omega_n}) + (j\frac{\omega}{\omega_n})^2$
- $= \left(1 - \frac{\omega^2}{\omega_n^2}\right) + j\left(\frac{2\zeta\omega}{\omega_n}\right)$
- $\lim_{\omega \rightarrow 0} G(j\omega) = 1 \angle 0$
- $\lim_{\omega \rightarrow \infty} G(j\omega) = \infty \angle 180$



GAIN MARGIN

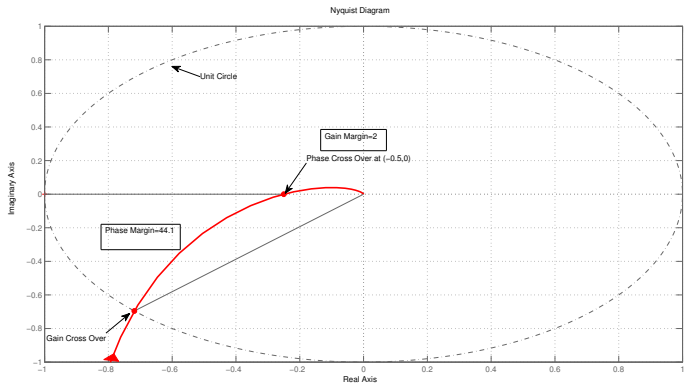
The gain margin is the reciprocal of the magnitude $|G(j\omega)|$ at the frequency at which the phase angle is -180° (Phase cross over frequency)

$$K_g = \frac{1}{|G(j\omega)|} \text{ at } \angle G = -180^\circ$$

PHASE MARGIN

The phase margin is the amount of additional phase lag at the gain cross over frequency required to the verge of instability $\gamma = 180^\circ + \phi$

PHASE AND GAIN MARGIN

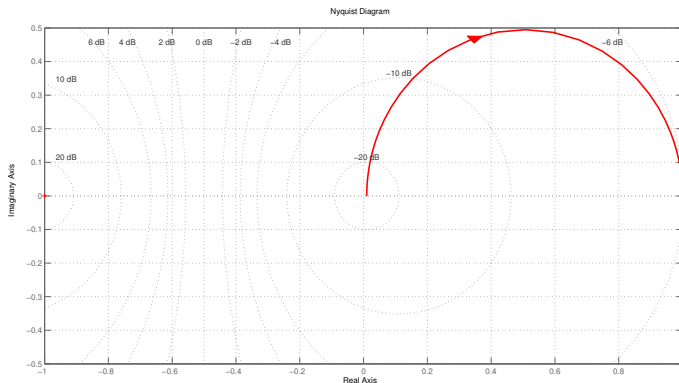


MORE EXAMPLE

$$G(s) = \frac{s + 1}{s + 100}$$

$$G(j\omega) = \frac{j\omega + 1}{j\omega + 100}$$

- $G(j0) = \frac{1}{100} \angle 0$ $G(j\infty) = 1 \angle 0$
- $\angle G = \tan^{-1} \omega - \tan^{-1} \frac{\omega}{100} > 0$

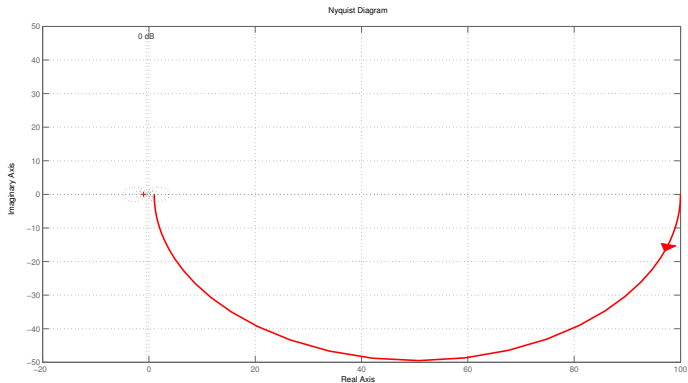


MORE EXAMPLE

$$G(s) = \frac{s + 100}{s + 1}$$

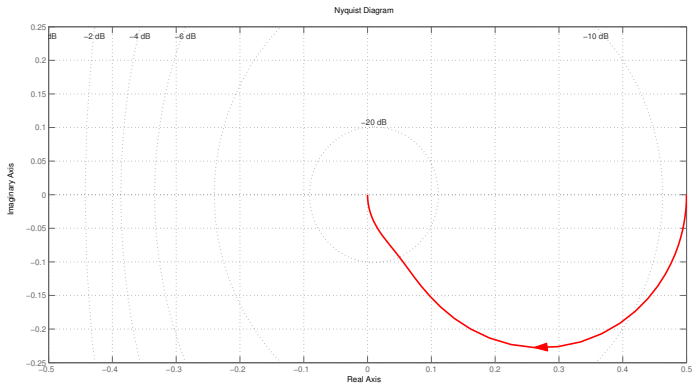
$$G(j\omega) = \frac{j\omega + 100}{j\omega + 1}$$

- $G(j0) = 100\angle 0$ $G(j\infty) = 1\angle 0$
- $\angle G = \tan^{-1} \frac{\omega}{100} - \tan^{-1} \omega < 0$



EXAMPLES

- $G(s) = \frac{s + 5}{(s + 1)(s + 10)}$
- $G(j0) = 0.5 \angle 0^\circ$ $G(j\infty) = 0 \angle -90^\circ$



EXAMPLE

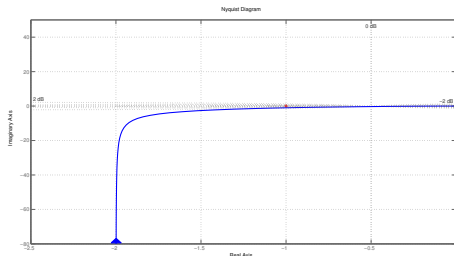
$$G(s) = \frac{1}{s(sT + 1)}$$

$$G(j\omega) = \frac{1}{j\omega(j\omega T + 1)} = -\frac{T}{1 + \omega^2 T^2} - j\frac{1}{\omega(1 + \omega^2 T^2)}$$

$$\lim_{\omega \rightarrow 0} G(j\omega) = -T - j\infty = \infty \angle -90^\circ$$

$$\lim_{\omega \rightarrow \infty} G(j\omega) = 0 - j0 = 0 \angle -180^\circ$$

Suppose $G(s) = \frac{1}{s(5s + 1)}$



EXAMPLE

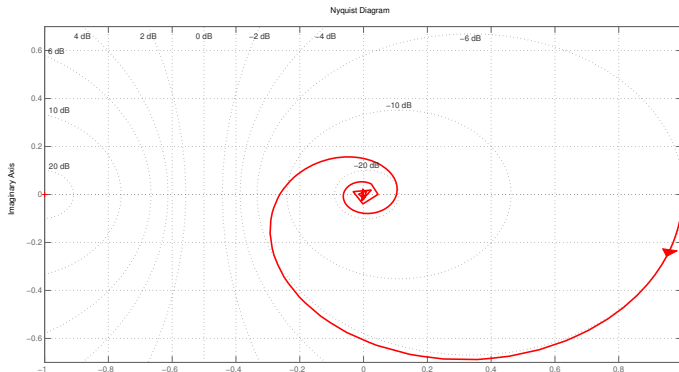
$$G(s) = \frac{e^{-s}}{2s + 1}$$

$$G(j\omega) = \frac{e^{-j\omega}}{2j\omega + 1}$$

Magnitude = $\frac{1}{\sqrt{4\omega^2 + 1}}$ Magnitude decreases with frequency

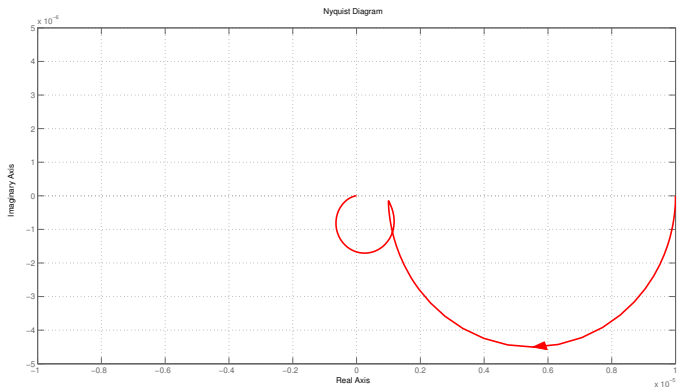
Phase = $-\omega - \tan^{-1} 4\omega$ Phase is changes with frequency continuously.

Polar plot starts from $1\angle 0$. Its spiral in nature.



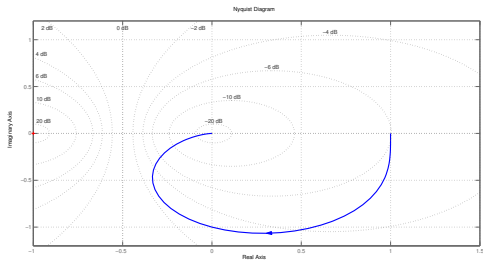
MORE EXAMPLE

- $G(s) = \frac{s + 10}{(s + 1)(s^2 + 600s + 1000000)}$
- $G(j\omega) = 10^{-5} \angle 0^\circ \quad G(j\infty) = 0 \angle -180^\circ$



$$G(s) = \frac{(s + z_1)(s + z_2) \dots (s + z_m)}{s^k(s + p_1)(s + p_2) \dots (s + p_n)} \quad n > m$$

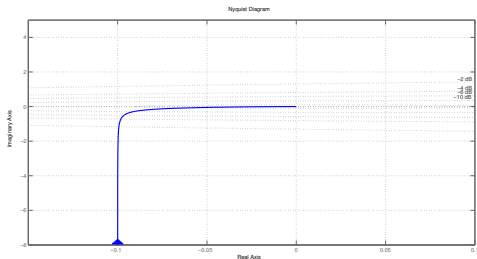
- Type 0 System:
Starting point($\omega = 0$) is finite and on real axis.
Ending point($\omega = \infty$) is at the origin.



SYSTEM TYPE AND POLAR PLOT

$$G(s) = \frac{(s + z_1)(s + z_2) \dots (s + z_m)}{s^k (s + p_1)(s + p_2) \dots (s + p_n)} \quad n > m$$

- Type 1 System: At $\omega = 0$ the magnitude is 0 and phase become -90° . At low frequency polar plot is asymptotic to a line parallel to imaginary axis. At $\omega = \infty$ magnitude become zero.



SYSTEM TYPE AND POLAR PLOT

$$G(s) = \frac{(s + z_1)(s + z_2) \dots (s + z_m)}{s^k (s + p_1)(s + p_2) \dots (s + p_n)} \quad n > m$$

- Type 2 System: At $\omega = 0$ the magnitude is ∞ but angle is -180° . So at lower frequency polar plot is asymptotic to a line parallel to negative real axis. At $\omega = \infty$ magnitude is zero.

